

### SGM6013 1.6MHz, 800mA Synchronous **Step-Down Converter**

### GENERAL DESCRIPTION

The SGM6013 is a 1.6MHz, constant frequency, current mode, synchronous, step-down switching regulator. It can deliver 800mA load current from 2.5V to 5.5V input voltage, and output voltage can be as low as 0.6V.

High switching frequency minimizes the sizes of inductor and capacitor. Integrated power MOSFETs and internal compensation make the SGM6013 simple to use and fit the total solution in a compact space.

The SGM6013 can also run at 100% duty cycle for low dropout operation, extending battery life in portable system. The synchronous architecture eliminates the external Schottky diode, and achieves over 90% of the power conversion efficiency. Low output ripple voltage at light load, 30µA quiescent current and less than 1µA of shutdown current make SGM6013 the ideal power supply solution for portable applications.

SGM6013 is available in both adjustable and fixed (1.2V, 1.8V, 3.3V) output voltage versions. It is in the Green TSOT-23-5 and TDFN-2×2-6L packages. It is rated over the -40°C to +85°C temperature range.

### TYPICAL APPLICATION



### **FEATURES**

- High Efficiency: Up to 95%
- 30µA Low Quiescent Current at Light Load •
- **800mA Output Current** •
- Input Voltage Range: 2.5V to 5.5V •
- 1.2V, 1.8V & 3.3V Fixed & Adjustable Output Voltages •
- 0.6V Reference Voltage •
- 1.6MHz Constant Switching Frequency
- Less than 1µA Shutdown Current •
- 100% Duty Cycle for Lowest Dropout
- No External Power MOSFETs and Schottky Diode Required
- **Excellent Line Regulation & Load Transient Response**
- -40°C to +85°C Operating Temperature Range
- Available in Green TSOT-23-5 and TDFN-2×2-6L Packages

### APPLICATIONS

GPS **MP3** Players Cellular, Smart Phones **Digital Book Readers Digital Still Cameras** Portable Instruments Wireless and DSL Modems **Battery Powered Equipments** Microprocessor, DSP Power Supplies



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### 1.6MHz, 800mA Synchronous Step-Down Converter

### **PACKAGE/ORDERING INFORMATION**

MODEL	V <sub>оит</sub> (V)	PIN- PACKAGE	SPECIFIED TEMPERATURE RANGE	ORDERING NUMBER	PACKAGE MARKING	PACKAGE OPTION
	1.2V	TSOT-23-5	-40°C to +85°C	SGM6013-1.2YTN5G/TR	SC1XX	Tape and Reel, 3000
	1.8V	TSOT-23-5	-40°C to +85°C	SGM6013-1.8YTN5G/TR	SH5XX	Tape and Reel, 3000
SGM6013	3.3V	TSOT-23-5	-40°C to +85°C	SGM6013-3.3YTN5G/TR	SH6XX	Tape and Reel, 3000
	Adjustable	TSOT-23-5	-40°C to +85°C	SGM6013-ADJYTN5G/TR	SC2XX	Tape and Reel, 3000
	Adjustable	TDFN-2×2-6L	-40°C to +85°C	SGM6013-ADJYTDI6G/TR	SC2 XXXX	Tape and Reel, 3000

NOTE: XX = Date Code. XXXX = Date Code.

Green (RoHS & HSF): SG Micro Corp defines "Green" to mean Pb-Free (RoHS compatible) and free of halogen substances. If you have additional comments or questions, please contact your SGMICRO representative directly.

### **ABSOLUTE MAXIMUM RATINGS**

Input Supply Voltage RUN, V <sub>FB</sub> Voltages SW Voltage Package Thermal Resistance	0.3V to $V_{\text{IN}}$
TSOT-23-5, θ <sub>JA</sub>	200°C/W
TDFN-2×2-6L, θ <sub>JA</sub>	200°C/W
P-Channel Switch Source Current (DC) N-Channel Switch Sink Current (DC) Peak SW Sink and Source Current Operating Temperature Range	800mA 1.3A
Junction Temperature	150°C
Storage Temperature Range	65°C to +150°C
Lead Temperature (Soldering, 10s)	260°C
ESD Susceptibility HBM MM	

#### MARKING INFORMATION



For example: SC1CD(2012, April)

#### NOTE:

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

### CAUTION

This integrated circuit can be damaged by ESD if you don't pay attention to ESD protection. SGMICRO recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage. ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

SGMICRO reserves the right to make any change in circuit design, specification or other related things if necessary without notice at any time. Please contact SGMICRO sales office to get the latest datasheet.



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### PIN CONFIGURATIONS (TOP VIEW)



### **PIN DESCRIPTION**

I	PIN TSOT-23-5 TDFN-2×2-6L		PIN		FUNCTION
TSOT-23-5			FUNCTION		
1	3	V <sub>IN</sub>	Supply Voltage Pin. Must be closely decoupled to GND, with a $4.7\mu\text{F}$ or greater ceramic capacitor.		
2	2, 4	GND	Ground.		
3	1	RUN	Run Control Input. Forcing this pin above 1.5V enables the part. Forcing this pin below 0.3V shuts down the device. In shutdown, all functions are disabled drawing <1µA supply current. Do not leave RUN floating.		
4	6	$V_{FB}$	Feedback Pin. Receives the feedback voltage from an external resistive divider across the output. The internal voltage divider is disabled for this adjustable version. (SGM6013-ADJ)		
4	_	V <sub>OUT</sub>	Output Voltage Feedback Pin. An internal resistive divider divides the output voltage down for comparison to the internal reference voltage. (SGM6013-1.2/SGM6013-1.8/SGM6013-3.3)		
5	5	SW	Switch Node Connection to Inductor. This pin connects to the drains of the internal main and synchronous power MOSFET switches.		



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### **ELECTRICAL CHARACTERISTICS**

 $(V_{IN} = 3.6V, L = 2.2 \mu H, C_{IN} = 4.7 \mu F, C_{OUT} = 10 \mu F, Full = -40^{\circ}C \text{ to } +85^{\circ}C, \text{ typical values are at } T_{A} = +25^{\circ}C, \text{ unless otherwise noted.})$ 

PARAMETER		SYMBOL	CONDI	TIONS	TEMP	MIN	TYP	MAX	UNITS
Input Voltage Range		V <sub>IN</sub>			Full	2.5		5.5	V
Feedback Currer	nt	I <sub>VFB</sub>			Full		±1	±100	nA
					Full	0.580	0.600	0.622	
Regulated Feedl	oack Voltage	V <sub>FB</sub>			+25°C	0.583	0.600	0.620	V
			$0^{\circ}C \le T_A \le +85^{\circ}C$	2		0.582	0.600	0.621	
Reference Voltag	ge Line Regulation	$\Delta V_{FB}$	V <sub>IN</sub> = 2.5V to 5.5	νV	Full		0.1	0.6	%/V
			SGM6013-1.2	I <sub>OUT</sub> = 100mA		1.159	1.200	1.241	
Regulated Outpu	it Voltage	V <sub>OUT</sub>	SGM6013-1.8	I <sub>OUT</sub> = 100mA	Full	1.739	1.800	1.861	V
			SGM6013-3.3	I <sub>OUT</sub> = 100mA	_	3.188	3.300	3.412	
Output Voltage Line Regulation		$\Delta V_{OUT}$	V <sub>IN</sub> = 2.5V to 5.5V		Full		0.1	0.6	%/V
Peak Inductor Current		I <sub>PK</sub>	$V_{FB}$ = 0.5V or $V_{OUT}$ = 90%, $V_{IN}$ = 3V		+25°C		1	1.25	Α
Output Voltage Load Regulation		VLOADREG			+25°C		0.5		%
SW Leakage Cu	rrent	I <sub>sw</sub>	$V_{RUN} = 0V, V_{SW} = 0V \text{ or } 5V,$ $V_{IN} = 5V$		+25°C		±0.01	±1	μA
	PWM Mode		$V_{FB} = 0.5V \text{ or } V_{OUT} = 90\%,$ $I_{LOAD} = 0A$				280	360	
Supply Current	PFM Mode	Is	$V_{FB} = 0.62V \text{ or } V$ $I_{LOAD} = 0A$	/ <sub>OUT</sub> = 103%,	+25℃		30	56	μA
	Shutdown		$V_{RUN} = 0V, V_{IN} =$	4.2V			0.1	1	
RUN Threshold		Vін			Full	1.5			v
RUN ITTESTOR		VIL			Full			0.3	v
RUN Leakage Current		I <sub>RUN</sub>			Full		±0.01	±1	μA
0 11 4 5		£	V <sub>FB</sub> = 0.6V or V <sub>OUT</sub> = 100%		Full	1.3	1.6	1.9	MHz
Oscillator Frequency		f <sub>osc</sub>	V <sub>FB</sub> = 0V or V <sub>OUT</sub> = 0V		+25°C		200		kHz
R <sub>DS(ON)</sub> of P-Channel FET		R <sub>PFET</sub>	I <sub>SW</sub> = 100mA		+25°C		0.46	0.65	Ω
R <sub>DS(ON)</sub> of N-Cha	nnel FET	R <sub>NFET</sub>	I <sub>SW</sub> = -100mA		+25°C		0.36	0.56	Ω
PFM/PWM Mode	e Switch Point				+25°C		40		mA



## 1.6MHz, 800mA Synchronous Step-Down Converter

### **TYPICAL PERFORMANCE CHARACTERISTICS**

 $T_A$  = +25°C, L = 2.2µH,  $C_{IN}$  = 4.7µF,  $C_{OUT}$  = 10µF, unless otherwise noted.



## 1.6MHz, 800mA Synchronous Step-Down Converter

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 $T_{A}$  = +25°C, L = 2.2µH,  $C_{IN}$  = 4.7µF,  $C_{OUT}$  = 10µF, unless otherwise noted.



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RUN

Vout

 $I_{L}$ 

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### **TYPICAL PERFORMANCE CHARACTERISTICS**

1V/div

500mA/div



 $T_{A}$  = +25°C, L = 2.2µH,  $C_{IN}$  = 4.7µF,  $C_{OUT}$  = 10µF, unless otherwise noted.

 $V_{IN} = 3.6V, V_{OUT} = 1.8V, I_{LOAD} = 600 \text{mA}$ 

Time (400µs/div)



### **TYPICAL APPLICATION CIRCUITS**



Figure 1. Single Li-ion 1.2V/800mA Regulator for High Efficiency and Small Footprint



Figure 2. Single Li-ion 1.8V/800mA Regulator for High Efficiency and Small Footprint



Figure 3. Tiny 3.3V/800mA Buck Regulator for High Efficiency and Small Footprint

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### **TYPICAL APPLICATION CIRCUITS**



Figure 4. Basic Application Circuit with Adjustable Version

V <sub>IN</sub>	V <sub>out</sub> (V)	С <sub>оит</sub> (µF)	C1 (pF)	R1 (kΩ)	R2 (kΩ)
2.5V to 5.5V	1.2	22	18	300	300
2.5V to 5.5V	1.3	22	18	301	261
2.5V to 5.5V	1.5	22	18	300	200
2.5V to 5.5V	1.8	10	18	300	150
$V_{OUT}$ to 5.5V	2.5	10	18	301	95.3
$V_{\text{OUT}}$ to 5.5V	2.7	10	18	301	86.6
$V_{\text{OUT}}$ to 5.5V	2.8	10	18	301	82.5
$V_{\text{OUT}}$ to 5.5V	3	10	18	300	75
$V_{OUT}$ to 5.5V	3.1	10	18	300	72
$V_{OUT}$ to 5.5V	3.3	10	18	300	66.5

NOTE: It is recommended to use a ceramic capacitor (C1) to obtain the best load transient response.



### **APPLICATION INFORMATION**

The basic SGM6013 application circuits are shown in Figure 1, 2, 3 and 4. External component selection is driven by the load requirement and begins with the selection of L followed by  $C_{\rm IN}$  and  $C_{\rm OUT}.$ 

#### **Inductor Selection**

For most applications, the value of the inductor will fall in the range of 1µH to 4.7µH. Its value is chosen based on the desired ripple current. Large value inductors lower ripple current and small value inductors result in higher ripple currents. Higher  $V_{IN}$  or  $V_{OUT}$  also increases the ripple current as shown in equation 1. A reasonable starting point for setting ripple current is  $\Delta I_L = 240$ mA (30% of 800mA).

$$\Delta I_{L} = \frac{1}{(f)(L)} V_{OUT} \left( 1 - \frac{V_{OUT}}{V_{IN}} \right)$$
(1)

The DC current rating of the inductor should be at least equal to the maximum load current plus half the ripple current to prevent core saturation. Thus, a 920mA rated inductor should be enough for most applications (800mA + 120mA). For better efficiency, choose a low DC-resistance inductor.

The inductor value also has an effect on power saving mode operation. The transition to low current operation begins when the inductor current peaks fall to approximately 100mA. Lower inductor values (higher  $\Delta I_L$ ) will cause this to occur at lower load currents, which can cause a dip in efficiency in the upper range of low current operation. In power saving mode operation, lower inductance values will cause the burst frequency to increase.

#### **Inductor Core Selection**

Different core materials and shapes will change the size/current and price/current relationship of an inductor. Toroid or shielded pot cores in ferrite or permalloy materials are small and don't radiate much energy, but generally cost more than powdered iron core inductors with similar electrical characteristics. The choice of which style inductor to use often depends more on the price vs size requirements and any radiated field/EMI requirements than on what the SGM6013 requires to operate.

#### C<sub>IN</sub> and C<sub>OUT</sub> Selection

In continuous mode, the source current of the top MOSFET is a square wave of duty cycle  $V_{OUT}/V_{IN}$ . To prevent large voltage transients, a low ESR input capacitor sized for the maximum RMS current must be used. The maximum RMS capacitor current is given by:

$$C_{\text{IN}} \text{ required } I_{\text{RMS}} \cong I_{\text{OMAX}} \frac{\left[V_{\text{OUT}} \left(V_{\text{IN}} - V_{\text{OUT}}\right)\right]^{1/2}}{V_{\text{IN}}}$$

This formula has a maximum at  $V_{IN} = 2V_{OUT}$ , where  $I_{RMS} = I_{OUT}/2$ . This simple worst-case condition is commonly used for design because even significant deviations do not offer much relief. Note that the capacitor manufacturer's ripple current ratings are often based on 2000 hours of life. This makes it advisable to further derate the capacitor, or choose a capacitor rated at a higher temperature than required. Always consult the manufacturer if there is any question.

The selection of  $C_{\text{OUT}}$  is driven by the required effective series resistance (ESR).

Typically, once the ESR requirement for  $C_{OUT}$  has been met, the RMS current rating generally far exceeds the  $I_{RIPPLE(P-P)}$  requirement. The output ripple  $\Delta V_{OUT}$  is determined by:

$$\Delta V_{\text{OUT}} \cong \Delta I_{\text{L}} \left( \text{ESR} + \frac{1}{8 \text{fC}_{\text{OUT}}} \right)$$

where f = operating frequency,  $C_{OUT}$  = output capacitance and  $\Delta I_L$  = ripple current in the inductor. For a fixed output voltage, the output ripple is highest at maximum input voltage since  $\Delta I_L$  increases with input voltage.

Aluminum electrolytic and dry tantalum capacitors are both available in surface mount configurations. In the case of tantalum, it is critical that the capacitors be surge tested for use in switching power supplies. An excellent choice is the AVX TPS series of surface mount tantalum. These are specially constructed and tested for low ESR so they give the lowest ESR for a given volume. Other capacitor types include Sanyo POSCAP, Kemet T510 and T495 series.



### **APPLICATION INFORMATION**

#### **Using Ceramic Input and Output Capacitors**

Higher values, lower cost ceramic capacitors are now becoming available in smaller case sizes. Their high ripple current, high voltage rating and low ESR make them ideal for switching regulator applications. Because the SGM6013's control loop does not depend on the output capacitor's ESR for stable operation, ceramic capacitors can be used freely to achieve very low output ripple and small circuit size.

However, care must be taken when ceramic capacitors are used at the input and the output. When a ceramic capacitor is used at the input and the power is supplied by a wall adapter through long wires, a load step at the output can induce ringing at the input,  $V_{\rm IN}$ . At best, this ringing can couple to the output and be mistaken as loop instability. At worst, a sudden inrush of current through the long wires can potentially cause a voltage spike at  $V_{\rm IN}$ , large enough to damage the part.

When choosing the input and output ceramic capacitors, choose the X5R or X7R dielectric formulations. These dielectrics have the best temperature and voltage characteristics of all the ceramics for a given value and size.

#### **Output Voltage Programming**

In the adjustable version, the output voltage is set by a resistive divider according to the following formula:

$$V_{\text{OUT}} = 0.6V \left(1 + \frac{\text{Rl}}{\text{R2}}\right)$$
(2)

The external resistive divider is connected to the output, allowing remote voltage sensing as shown in Figure 5.



Figure 5. Setting the SGM6013 Output Voltage

#### **Thermal Considerations**

In most applications the SGM6013 does not dissipate much heat due to its high efficiency. But, in applications where the SGM6013 is running at high ambient temperature with low supply voltage and high duty cycles, such as in dropout, the heat dissipated may exceed the maximum junction temperature of the part. If the junction temperature reaches approximately 150°C, both power switches will be turned off and the SW node will become high impedance.

To avoid the SGM6013 from exceeding the maximum junction temperature, the user will need to do some thermal analysis. The goal of the thermal analysis is to determine whether the power dissipated exceeds the maximum junction temperature of the part. The temperature rise is given by:

$$T_{R} = (P_{D})(\theta_{JA})$$

where  $P_D$  is the power dissipated by the regulator and  $\theta_{JA}$  is the thermal resistance from the junction of the die to the ambient temperature. The junction temperature,  $T_J$ , is given by:  $T_J = T_A + T_R$  where  $T_A$  is the ambient temperature.



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### **APPLICATION INFORMATION**

#### PC Board Layout Guidelines

When laying out the printed circuit board, the following guidelines should be used to ensure proper operation of the SGM6013. These items are also illustrated graphically in Figures 6. Check the following in your layout:

1. Keep the traces of  $V_{IN}$ , GND and SW short and wide.

2. Connect the V<sub>FB</sub> pin directly to the feedback resistors. The resistive divider R2/R1 must be connected between the (+) plate of  $C_{OUT}$  and ground.

3. Connect the (+) plate of  $C_{IN}$  as closely to  $V_{IN}$  as possible.

4. Keep the switching node, SW, away from the sensitive  $V_{\text{FB}}$  node.

5. Keep the (–) plates of  $C_{\text{IN}}$  and  $C_{\text{OUT}}$  as close as possible.



Figure 6a. SGM6013 Layout Diagram (TSOT-23-5)



Figure 6b. SGM6013-1.8 Layout Diagram (TSOT-23-5)



### 1.6MHz, 800mA Synchronous Step-Down Converter

### **APPLICATION INFORMATION**







Figure 7b

Figure 7 shows the complete circuit along with its efficiency curve.



### Single Li-ion 1.2V/800mA Regulator for High Efficiency and Small Footprint



#### Single Li-ion 1.8V/800mA Regulator for High Efficiency and Small Footprint











### Tiny 3.3V/800mA Buck Regulator for High Efficiency and Small Footprint





#### **Basic Application Circuit with Adjustable Version**









### 1.6MHz, 800mA Synchronous Step-Down Converter

## PACKAGE OUTLINE DIMENSIONS

**TSOT-23-5** 





#### RECOMMENDED LAND PATTERN (Unit: mm)





Symbol		nsions imeters	Dimensions In Inches			
	MIN	MAX	MIN	MAX		
A	0.700	0.900	0.028	0.035		
A1	0.000	0.100	0.000	0.004		
A2	0.700	0.800	0.028	0.031		
b	0.350	0.500	0.014	0.020		
с	0.080	0.200	0.003	0.008		
D	2.820	3.020	0.111	0.119		
E	1.600	1.700	0.063	0.067		
E1	2.650	2.950	0.104	0.116		
e	0.950	) BSC	0.037 BSC			
e1	1.900 BSC		0.075	BSC		
L	0.300	0.600	0.012	0.024		
θ	0°	8°	0°	8°		



### PACKAGE OUTLINE DIMENSIONS

TDFN-2×2-6L



SIDE VIEW



N6

Ā E1

N1

RECOMMENDED LAND PATTERN (Unit: mm)

0.65

Symbol	-	nsions meters	Dimensions In Inches			
	MIN	MAX	MIN	MAX		
A	0.700	0.800	0.028	0.031		
A1	0.000	0.050	0.000	0.002		
A2	0.203	3 REF	0.008 REF			
D	1.900	2.100	0.075	0.083		
D1	1.100	1.300	0.043	0.051		
E	1.900	2.100	0.075	0.083		
E1	0.600	0.800	0.024	0.031		
k	0.200	) MIN	0.008	8 MIN		
b	0.180	0.300	0.007	0.012		
е	0.650	) TYP	0.026	TYP		
L	0.250	0.450	0.010	0.018		

0.24



### TAPE AND REEL INFORMATION

### **REEL DIMENSIONS**



NOTE: The picture is only for reference. Please make the object as the standard.

#### KEY PARAMETER LIST OF TAPE AND REEL

Package Type	Reel Diameter	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P0 (mm)	P1 (mm)	P2 (mm)	W (mm)	Pin1 Quadrant
TSOT-23-5	7"	9.5	3.17	3.1	1.10	4.0	4.0	2.0	8.0	Q3
TDFN-2×2-6L	7"	9.5	2.30	2.30	1.10	4.00	4.00	2.00	8.00	Q1



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#### **CARTON BOX DIMENSIONS**



NOTE: The picture is only for reference. Please make the object as the standard.

#### **KEY PARAMETER LIST OF CARTON BOX**

Reel Type	Length (mm)			Pizza/Carton
7" (Option)	368	227	224	8
7"	442	410	224	18

