# OPAx340 Single-Supply, Rail-to-Rail Operational Amplifiers MicroAmplifier™ Series

#### 1 Features

Rail-to-Rail Input

Rail-to-Rail Output (Within 1 mV)

MicroSize Packages

Wide Bandwidth: 5.5 MHzHigh Slew Rate: 6 V/µs

Low THD + Noise: 0.0007% (f = 1 kHz)
Low Quiescent Current: 750 μA/Channel

Single, Dual, and Quad Versions

### 2 Applications

Driving A/D Converters

PCMCIA Cards

Data Acquisition

Process Control

· Audio Processing

Communications

Active Filters

Test Equipment

### 3 Description

The OPA340 series rail-to-rail CMOS operational amplifiers are optimized for low-voltage, single-supply operation. Rail-to-rail input and output and high-speed operation make them ideal for driving sampling analog-to-digital (A/D) converters. They are also well-suited for general purpose and audio applications as well as providing I/V conversion at the output of digital-to-analog (D/A) converters. Single, dual, and quad versions have identical specifications for design flexibility.

The OPA340 series operate on a single supply as low as 2.5 V with an input common-mode voltage range that extends 500 mV below ground and 500 mV above the positive supply. Output voltage swing is to within 1 mV of the supply rails with a 100-k $\Omega$  load. These devices offer excellent dynamic response (BW = 5.5 MHz, SR = 6 V/ $\mu$ s), yet quiescent current is only 750 A. Dual and quad designs feature completely independent circuitry for lowest crosstalk and freedom from interaction.

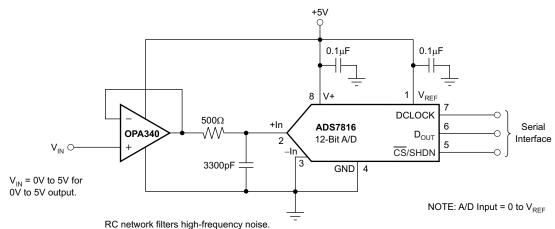
The single (OPA340) packages are the tiny 5-pin SOT-23 surface mount, 8-pin SOIC surface mount, and 8-pin DIP. The dual (OPA2340) comes in the miniature 8-pin VSSOP surface mount, 8-pin SOIC surface mount, and 8-pin PDIP packages. The quad (OPA4340) packages are the space-saving 16-pin SSOP surface mount and 14-pin SOIC surface mount. All are specified from -40°C to 85°C and operate from -55°C to 125°C. A SPICE macromodel is available for design analysis.

#### Device Information<sup>(1)</sup>

PART NUMBER	PACKAGE	BODY SIZE (NOM)
OPA340	SOT-23 (5)	3.00 mm × 3.00 mm
OPA340, OPA2340	PDIP (8)	9.81 mm × 6.35 mm
OPA340, OPA2340	SOIC (8)	4.90 mm × 3.91 mm
OPA2340	VSSOP (8)	3.00 mm × 3.00 mm
OPA4340	SOIC (14)	8.65 mm × 3.91 mm
UPA4340	SSOP (16)	4.90 mm × 3.90 mm

(1) For all available packages, see the orderable addendum at the end of the data sheet.

### **OPA340 in Noninverting Configuration Driving ADS7816**



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### 4 Revision History

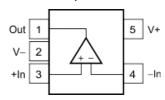
### Changes from Revision B (November 2007) to Revision C

**Page** 

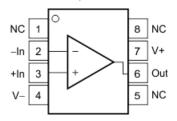
Added ESD Ratings table, Feature Description section, Device Functional Modes, Application and Implementation section, Power Supply Recommendations section, Layout section, Device and Documentation Support section, and Mechanical, Packaging, and Orderable Information section.
 Deleted Package/Ordering Information table, see POA at the end of the data sheet.

# 5 Pin Configuration and Functions

OPA340: DBV Package 5-Pin SOT-23 Top View



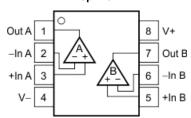
#### OPA340: P and D Packages 8-Pin PDIP and SOIC Top View



#### **Pin Functions: OPA340**

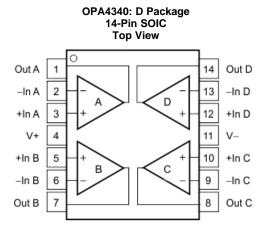
	PIN		I/O	DESCRIPTION	
NAME	SOT-23	SOIC, PDIP	1,0		
-IN	4	2	I	Negative (inverting) input	
+IN	3	3	I	Positive (noninverting) input	
NC	_	1, 5, 8	_	No internal connection (can be left floating)	
OUT	1	6	0	Output	
V-	2	4	_	Negative (lowest) power supply	
V+	5	7	_	Positive (highest) power supply	

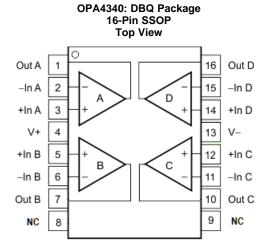
#### OPA2340: P, D, and DGK Packages 8-Pin PDIP, SOIC, and VSSOP Top View



Pin Functions: OPA2340

	NAME VSSOP, SOIC, PDIP		DESCRIPTION	
NAME				
-IN A	2	I	Negative (inverting) input channel A	
+IN A	3	I	Positive (noninverting) input channel A	
–IN B	6	I	Negative (inverting) input channel B	
+IN B	5	I	Positive (noninverting) input channel B	
OUT A	1	0	Output channel A	
OUT B	7	0	Output channel B	
V-	4	_	legative (lowest) power supply	
V+	8	_	Positive (highest) power supply	





### Pin Functions: OPA4340

	PIN		1/0	DESCRIPTION	
NAME	SOIC	SSOP	I/O	DESCRIPTION	
–IN A	2	2	I	Negative (inverting) input channel A	
–IN B	6	6	I	Negative (inverting) input channel B	
–IN C	9	11	I	Negative (inverting) input channel C	
–IN D	13	15	I	Negative (inverting) input channel D	
+IN A	3	3	I	Positive (noninverting) input channel A	
+IN B	5	5	I	Positive (noninverting) input channel B	
+IN C	10	12	I	Positive (noninverting) input channel C	
+IN D	12	14	I	Positive (noninverting) input channel D	
NC	_	8, 9	_	No internal connection (can be left floating)	
OUT A	1	1	0	Output, channel A	
OUT B	7	7	0	Output, channel B	
OUT C	8	10	0	Output, channel C	
OUT D	14	16	0	Output, channel D	
V-	11	13		Negative (lowest) power supply	
V+	4	4	_	Positive (highest) power supply	

### 6 Specifications

#### 6.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)<sup>(1)</sup>

		MIN	MAX	UNIT	
Voltogo	Supply voltage		5.5	\/	
Voltage	Signal input terminals <sup>(2)</sup>	-0.5	0.5	V	
Current	Signal input terminals <sup>(2)</sup>		10	mA	
Current	Output short circuit <sup>(3)</sup>	Conti	nuous		
	Operating, T <sub>A</sub>	-55	125		
Temperature	Junction, T <sub>J</sub>		150	°C	
	Storage, T <sub>stg</sub>	-55	125		

<sup>(1)</sup> Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

### 6.2 ESD Ratings

			VALUE	UNIT
\/	Electrostatic	Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 (1)	±600	.,
V(ESD)	discharge	Charged-device model (CDM), per JEDEC specification JESD22-C101 (2)	±250	V

<sup>(1)</sup> JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

### 6.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

	MIN	MAX	UNIT
Supply voltage	2.7	5.5	V
Specified temperature	-40	125	°C

#### 6.4 Thermal Information – OPA340

		OPA340				
		DBV (SOT-23)	P (PDIP)	D (SOIC)	D (SOIC)	UNIT
		5 PINS	8 PINS	8 PINS	14 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	207.9	53.1	142	83.8	°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	71.2	42.5	90.2	70.7	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	36.0	30.3	82.5	59.5	°C/W
ΨЈТ	Junction-to-top characterization parameter	2.0	19.7	39.4	11.6	°C/W
ΨЈВ	Junction-to-board characterization parameter	35.2	30.2	82	37.7	°C/W
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	_	_	_	_	°C/W

For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report.

<sup>(2)</sup> Input terminals are diode-clamped to the power-supply rails. Input signals that can swing more than 0.5 V beyond the supply rails should be current limited to 10 mA or less.

<sup>(3)</sup> Short-circuit to ground, one amplifier per package.

<sup>(2)</sup> JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

### 6.5 Thermal Information - OPA2340

		OPA	2340	
	Junction-to-case (top) thermal resistance  Junction-to-board thermal resistance  Junction-to-top characterization parameter  Junction-to-board characterization parameter	D (SOIC)	DGK (VSSOP)	UNIT
		8 PINS	8 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	138.4	169.2	°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	89.5	62.8	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	78.6	89.8	°C/W
ΨЈТ	Junction-to-top characterization parameter	29.9	7.5	°C/W
ΨЈВ	Junction-to-board characterization parameter	78.1	88.2	°C/W
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	_	_	°C/W

<sup>(1)</sup> For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report.

### 6.6 Thermal Information - OPA4340

		OPA4340	
	THERMAL METRIC <sup>(1)</sup>	DBQ (SSOP)	UNIT
		16 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	115.8	°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	67	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	58.3	°C/W
ΤιΨ	Junction-to-top characterization parameter	19.9	°C/W
ΨЈВ	Junction-to-board characterization parameter	57.9	°C/W
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	_	°C/W

<sup>(1)</sup> For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report.

### 6.7 Electrical Characteristics

At  $T_A$  = 25°C,  $R_L$  = 10 k $\Omega$  connected to  $V_S/2$ , and  $V_{OUT}$  =  $V_S/2$ , unless otherwise noted.

	PARAMETE	R	TEST CONDITIONS	MIN	TYP <sup>(1)</sup>	MAX	UNIT
OFFSET	VOLTAGE						
Vos	Input offset voltage		V <sub>S</sub> = 5 V		±150	±500	μV
dV <sub>OS</sub> /dt	Input offset voltage vs ter	mperature	$T_A = -40^{\circ}\text{C to } 85^{\circ}\text{C}, \ V_S = 5 \text{ V}$		±2.5		μV/°C
	l		V <sub>S</sub> = 2.7 V to 5.5 V, V <sub>CM</sub> = 0 V		30	120	μV/V
PSRR	Input offset voltage vs power supply	Over temperature	$V_S = 2.7 \text{ V to } 5.5 \text{ V}, V_{CM} = 0 \text{ V},$ $T_A = -40^{\circ}\text{C to } 85^{\circ}\text{C}, V_S = 5 \text{ V}$			120	μV/°C
	Channel separation, DC				0.2		μV/V
INPUT B	IAS CURRENT						
	Input bias current				±0.2	±10	pA
I <sub>S</sub>		Over temperature	$T_A = -40^{\circ}\text{C to } 85^{\circ}\text{C}, \ V_S = 5 \text{ V}$			±60	
Ios	Input offset current				±0.2	±10	pA
NOISE							
	Input voltage noise		f = 0.1 kHz to 50 kHz		8		$\mu V_{RMS}$
e <sub>n</sub>	Input voltage noise densi	ty	f = 1 kHz		25		nV/√ <del>Hz</del>
i <sub>n</sub>	Current noise density		f = 1 kHz		3		fA/√Hz
INPUT V	OLTAGE RANGE						
V <sub>CM</sub>	Common-mode voltage r	ange		-0.3		(V+) + 0.3	V
			-0.3 V < V <sub>CM</sub> < (V+) - 1.8 V	80	92		
CMRR	Common-mode rejection	ratio	V <sub>S</sub> = 5 V, -0.3 V < V <sub>CM</sub> < 5.3 V	70	84		dB
			V <sub>S</sub> = 2.7 V, -0.3 V < V <sub>CM</sub> < 3 V	66	80		

(1)  $V_S = 5 V$ .

# **Electrical Characteristics (continued)**

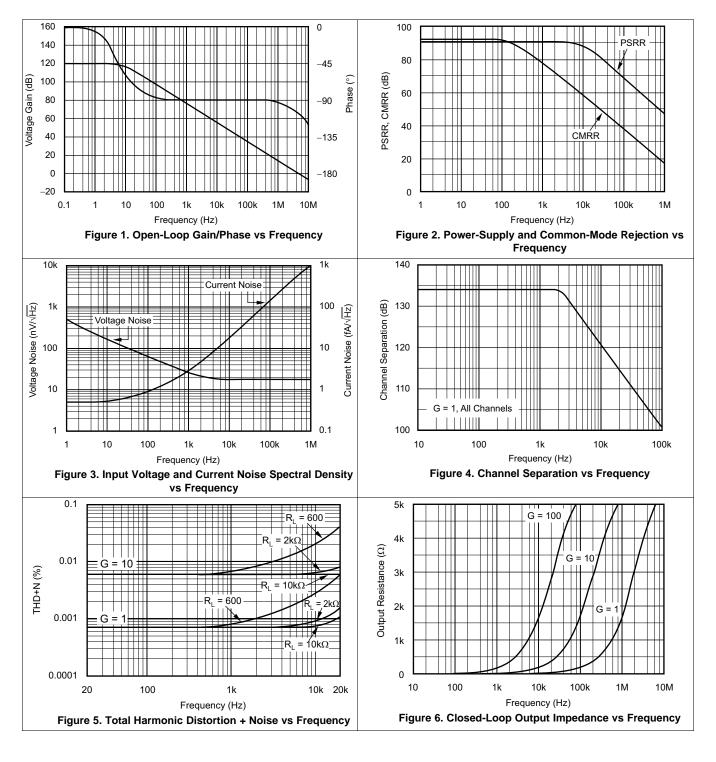
At T<sub>A</sub> = 25°C, R<sub>L</sub> = 10 k $\Omega$  connected to V<sub>S</sub>/2, and V<sub>OUT</sub> = V<sub>S</sub>/2, unless otherwise noted.

	PARAMETER		TEST CONDITIONS	MIN	TYP <sup>(1)</sup>	MAX	UNIT	
INPUT IN	MPEDANCE							
	Differential				10 <sup>13</sup>    3		$\Omega \parallel pF$	
	Common-mode				10 <sup>13</sup>    6		Ω    pF	
OPEN-LO	OOP GAIN							
			$R_L = 100 \text{ k}\Omega$ , 5 mV < $V_O$ < (V+) - 5 mV	106	124			
			$R_L = 10 \text{ k}\Omega,$ 5 mV < V <sub>O</sub> < (V+) - 50 mV	100	120			
			$R_L = 2 k\Omega$ , 200 mV < $V_O$ < (V+) - 200 mV	94	114			
A <sub>OL</sub>	Open-loop voltage gain		$R_L = 100 \text{ k}\Omega$ , $5 \text{ mV} < V_O < (V+) - 5 \text{ mV}$ , $T_A = -40^{\circ}\text{C to }85^{\circ}\text{C}$ , $V_S = 5 \text{ V}$	106			dB	
		Over temperature	$ \begin{aligned} R_L &= 10 \text{ k}\Omega, \\ 5 \text{ mV} &< V_O < (V+) - 50 \text{ mV}, \\ T_A &= -40^{\circ}\text{C to } 85^{\circ}\text{C}, V_S = 5 \text{ V} \end{aligned} $	100				
			$\begin{aligned} R_L &= 2 \ k\Omega, \\ 200 \ mV < V_O < (V+) - 200 \ mV, \\ T_A &= -40^{\circ} C \ to \ 85^{\circ} C, \ V_S = 5 \ V \end{aligned}$	94				
FREQUE	NCY RESPONSE							
GBW	Gain-bandwidth product		G = 1		5.5		MHz	
SR	Slew rate		$V_S = 5 \text{ V}, G = 1, C_L = 100 \text{ pF}$		6		V/µs	
	Settling time, 0.1%		$V_S = 5 \text{ V}, 2\text{-V step}, C_L = 100 \text{ pF}$		1		μs	
	Settling time, 0.01%		$V_S = 5 \text{ V}, 2\text{-V step}, C_L = 100 \text{ pF}$		1.6		μs	
	Overload recovery time		$V_{IN} \times G = V_{S}$		0.2		μs	
THD+N	Total harmonic distortion +	noise	$V_S = 5 \text{ V}, V_O = 3V_{PP}^{(2)}, G = 1,$ f = 1 kHz		0.0007%			
OUTPUT								
			$R_L = 100 \text{ k}\Omega, A_{OL} \ge 106 \text{ dB}$		1	5		
			$R_L = 10 \text{ k}\Omega, A_{OL} \ge 106 \text{ dB}$		10			
			$R_L = 2 \text{ k}\Omega, A_{OL} \ge 106 \text{ dB}$		40			
	Voltage output swing from rail (2)		$R_L = 100 \text{ k}\Omega, A_{OL} \ge 106 \text{ dB},$ $T_A = -40^{\circ}\text{C to } 85^{\circ}\text{C}, V_S = 5 \text{ V}$			5	mV	
		Over temperature	$R_L = 10 \text{ k}\Omega, A_{OL} \ge 106 \text{ dB},$ $T_A = -40^{\circ}\text{C to } 85^{\circ}\text{C}, V_S = 5 \text{ V}$			50		
			$R_L = 2 k\Omega$ , $A_{OL} \ge 106 dB$ , $T_A = -40^{\circ}C$ to 85°C, $V_S = 5 V$			200		
I <sub>SC</sub>	Short-circuit current				±50		mA	
C <sub>LOAD</sub>	Capacitive load drive			See Typ	ical Characterist	ics		
POWER	SUPPLY							
Vs	Specified voltage range			2.7		5	V	
	Operating voltage range		Lower end Higher end		2.5 5.5		V	
	Quiescent current		I <sub>O</sub> = 0, V <sub>S</sub> = 5 V	750		950	μA	
ΙQ	(per amplifier)	Over temperature	I <sub>O</sub> = 0, V <sub>S</sub> = 5 V, T <sub>A</sub> = -40°C to 85°C	-40°C to 85°C				
TEMPER	ATURE RANGE	•						
	Specified range			-40		85	°C	
	Operating range			-55		125	°C	
	Storage range			-55		125	°C	

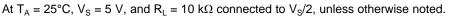
<sup>(2)</sup> Output voltage swings are measured between the output and power-supply rails.

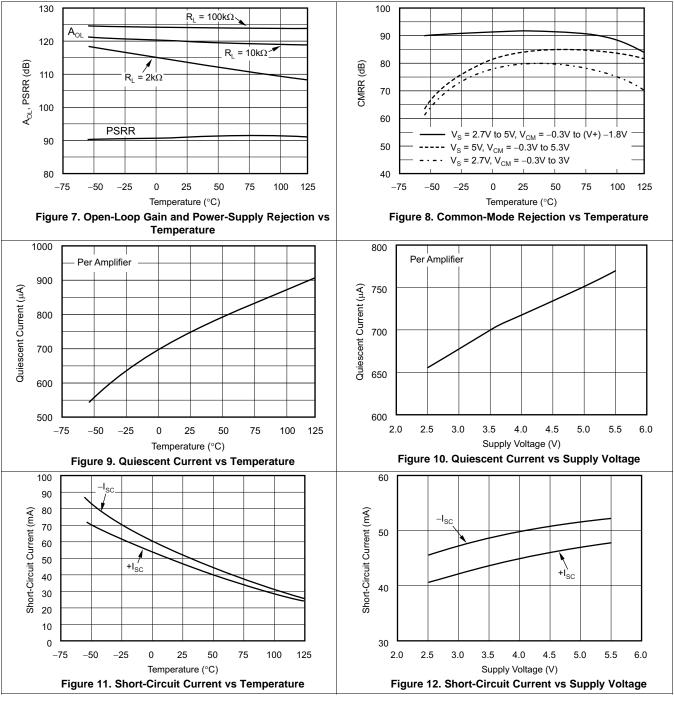
### 6.8 Typical Characteristics

At  $T_A$  = 25°C,  $V_S$  = 5 V, and  $R_L$  = 10 k $\Omega$  connected to  $V_S$ /2, unless otherwise noted.



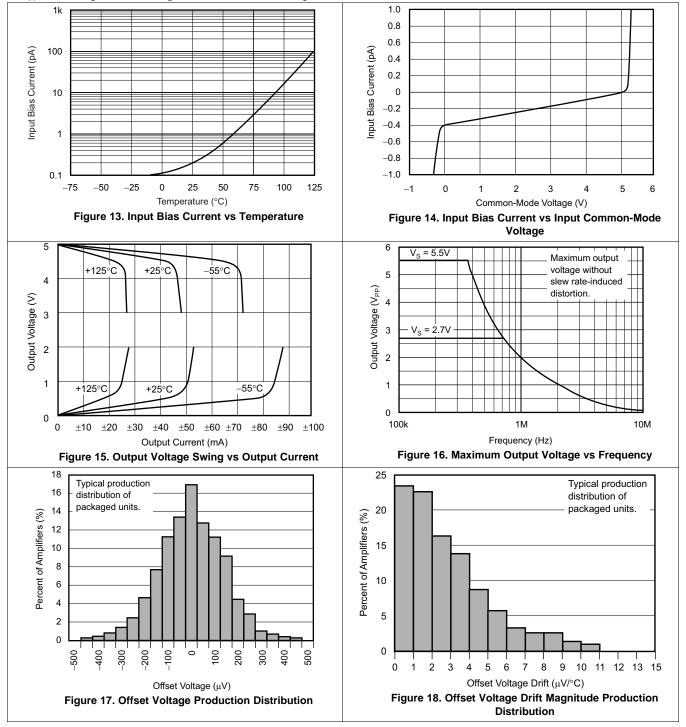
### **Typical Characteristics (continued)**





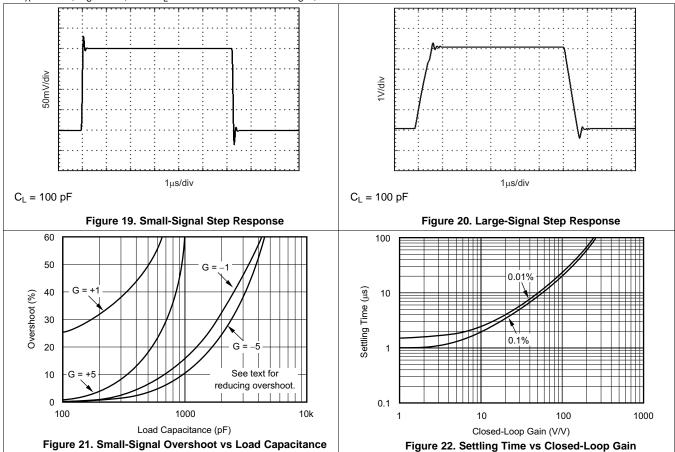
### **Typical Characteristics (continued)**

At  $T_A = 25$ °C,  $V_S = 5$  V, and  $R_L = 10$  k $\Omega$  connected to  $V_S/2$ , unless otherwise noted.



# **Typical Characteristics (continued)**

At  $T_A$  = 25°C,  $V_S$  = 5 V, and  $R_L$  = 10 k $\Omega$  connected to  $V_S/2$ , unless otherwise noted.



### 7 Detailed Description

#### 7.1 Overview

The OPA340 series operational amplifiers are fabricated on a state-of-the-art, 0.6-micron CMOS process. These devices are unity-gain stable and suitable for a wide range of general-purpose applications. Rail-to-rail input and output make them ideal for driving sampling A/D converters. In addition, excellent AC performance makes them well-suited for audio applications. The class AB output stage is capable of driving  $600-\Omega$  loads series and extends 500 mV beyond the supply. Rail-to-rail input and output swing significantly increases dynamic range, especially in low-supply applications. Figure 23 shows the input and output waveforms for the OPA340 in unity-gain configuration. Operation is from a single 5-V supply with a  $10-k\Omega$  load connected to V/2. The input is a  $5-V_{PP}$  sinusoid. Output voltage is approximately  $4.98~V_{PP}$ . Power-supply pins must be bypassed with  $0.01-\mu F$  ceramic capacitors.

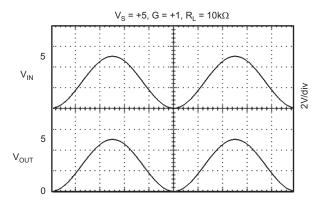
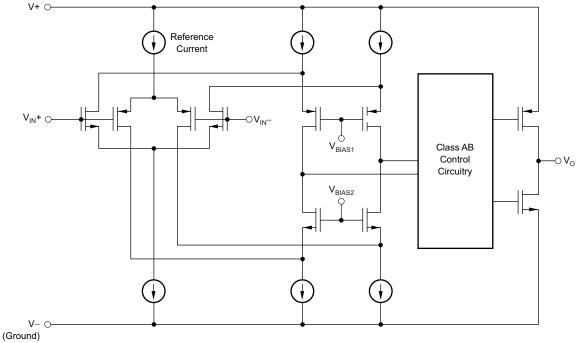


Figure 23. Rail-to-Rail Input and Output

#### 7.2 Functional Block Diagram



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#### 7.3 Feature Description

#### 7.3.1 Operating Voltage

The OPA340 series operational amplifiers are fully specified from 2.7 V to 5 V. However, supply voltage may range from 2.5 V to 5.5 V. Parameters are ensured over the specified supply range—a unique feature of the OPA340 series. In addition, many specifications apply from –40°C to 85°C. Most behavior remains virtually unchanged throughout the full operating voltage range. Parameters which vary significantly with operating voltages or temperature are shown in the *Typical Characteristics*.

#### 7.3.2 Rail-to-Rail Input

The input common-mode voltage range of the OPA340 series extends 500 mV beyond the supply rails. This extended range is achieved with a complementary input stage—an N-channel input differential pair in parallel with a P-channel differential pair. The N-channel pair is active for input voltages close to the positive rail, typically (V+) - 1.3 V to 500 mV above the positive supply, while the P-channel pair is on for inputs from 500 mV below the negative supply to approximately (V+) - 1.3 V. There is a small transition region, typically (V+) - 1.5 V to (V+) - 1.1 V, in which both pairs are on. This 400-mV transition region can vary  $\pm 300 \text{ mV}$  with process variation. Thus, the transition region (both stages on) can range from (V+) - 1.8 V to (V+) - 1.4 V on the low end, up to (V+) - 1.2 V to (V+) - 0.8 V on the high end.

OPA340 series operational amplifiers are laser-trimmed to the reduce offset voltage difference between the N-channel and P-channel input stages, resulting in improved common-mode rejection and a smooth transition between the N-channel pair and the P-channel pair. However, within the 400-mV transition region PSRR, CMRR, offset voltage, offset drift, and THD may be degraded compared to operation outside this region.

A double-folded cascode adds the signal from the two input pairs and presents a differential signal to the class AB output stage. Normally, input bias current is approximately 200 fA; however, input voltages exceeding the power supplies by more than 500 mV can cause excessive current to flow in or out of the input pins. Momentary voltages greater than 500 mV beyond the power supply can be tolerated if the current on the input pins is limited to 10 mA. This current limiting is easily accomplished with an input resistor, as shown in Figure 24. Many input signals are inherently current-limited to less than 10 mA; therefore, a limiting resistor is not required.

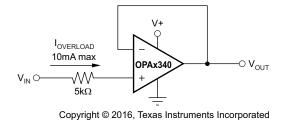


Figure 24. Input Current Protection for Voltages Exceeding the Supply Voltage

#### 7.3.3 Rail-to-Rail Output

A class AB output stage with common-source transistors is used to achieve rail-to-rail output. For light resistive loads (> 50 k $\Omega$ ), the output voltage is typically a few millivolts from the supply rails. With moderate resistive loads (2 k $\Omega$  to 50 k $\Omega$ ), the output can swing to within a few tens of millivolts from the supply rails and maintain high open-loop gain (see Figure 15).

#### 7.3.4 Capacitive Load and Stability

OPA340 series operational amplifiers can drive a wide range of capacitive loads. However, all operational amplifiers under certain conditions can become unstable, operational amplifier configuration, gain, and load value are some of the factors to consider when determining stability. An operational amplifier in unity-gain configuration is most susceptible to the effects of capacitive load. The capacitive load reacts with the output resistance of the operational amplifier, along with any additional load resistance, to create a pole in the small-signal response that degrades the phase margin. In unity-gain configuration, the OPA340 series operational amplifiers perform well, with a pure capacitive load up to approximately 1000 pF. Increasing gain enhances the amplifier ability to drive more capacitance (see Figure 21).

#### **Feature Description (continued)**

One method of improving capacitive load drive in the unity-gain configuration is to insert a  $10-\Omega$  to  $20-\Omega$  resistor in series with the output, as shown in Figure 25. This resistor significantly reduces ringing with large capacitive loads. However, if there is a resistive load in parallel with the capacitive load, it creates a voltage divider introducing a DC error at the output and slightly reduces output swing. This error can be insignificant. For instance, with  $R_L = 10~k\Omega$  and  $R_S = 20~\Omega$ , there is only an approximate 0.2% error at the output.

When used with the miniature package options of the OPA340 series, the combination is ideal for space-limited and low-power applications. For further information, consult the ADS7816 data sheet, 12-Bit High Speed Micro Power Sampling Analog-To-Digital Converter (SBAS061). With the OPA340 in a noninverting configuration, an RC network at the output of the amplifier can be used to filter high-frequency noise in the signal (see Figure 26). In the inverting configuration, filtering may be accomplished with a capacitor across the feedback resistor (see Figure 27).

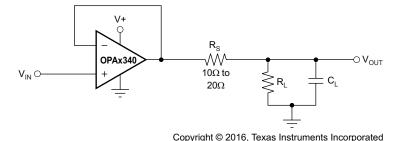


Figure 25. Series Resistor in Unity-Gain Configuration Improves Capacitive Load Drive

#### 7.3.5 Driving A/D Converters

The OPA340 series operational amplifiers are optimized for driving medium-speed (up to 100 kHz) sampling A/D converters. However, they also offer excellent performance for higher speed converters. The OPA340 series provides an effective means of buffering the converter input capacitance and resulting charge injection while providing signal gain. Figure 26 and Figure 27 show the OPA340 driving an ADS7816. The ADS7816 is a 12-bit, micro-power sampling converter in the tiny 8-pin VSSOP package.

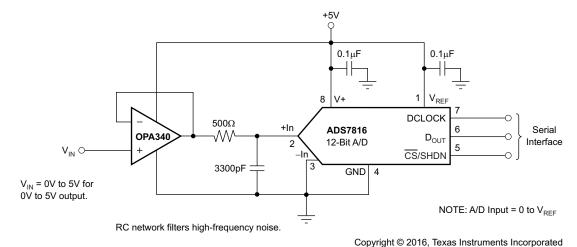


Figure 26. OPA340 in Noninverting Configuration Driving ADS7816

### **Feature Description (continued)**

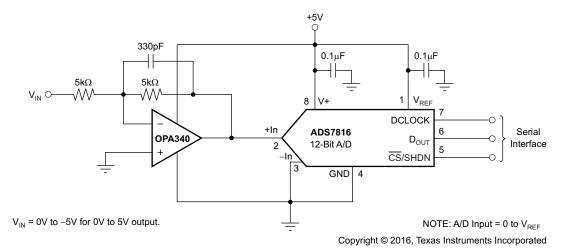


Figure 27. OPA340 in Inverting Configuration Driving ADS7816

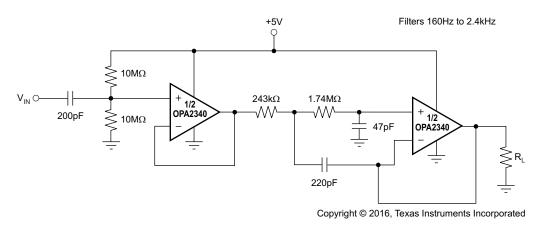


Figure 28. Speech Bandpass Filter

#### 7.4 Device Functional Modes

The OPAx340 has a single functional mode and is operational when the power-supply voltage is greater than 2.7 V (±1.35 V). The maximum power supply voltage for the OPAx340 is 5.5 V (±2.75 V).

### 8 Application and Implementation

#### NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

### 8.1 Application Information

The OPAx340 amplifier is a single-supply, CMOS operational amplifier with 5.5-MHz unity-gain bandwidth and supply current of 950  $\mu$ A. Its performance is optimized for low-voltage (2.7 V to 5.5 V), single-supply applications, with its input common-mode voltage linear range extending 300 mV beyond the rails and the output voltage swing within 5 mV of either rail. The OPAx340 series features wide bandwidth and unity-gain stability with rail-to-rail input and output for increased dynamic range. Power-supply pins must be bypassed with 0.01- $\mu$ F ceramic capacitors.

### 8.2 Typical Applications

#### 8.2.1 Single-Pole, Low-Pass Filter

Figure 29 shows the OPA340 in a typical noninverting application with the input signal bandwidth limited by the input lowpass filter.

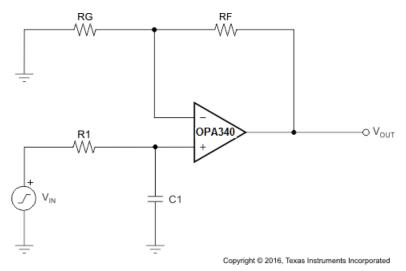


Figure 29. Single-Pole, Low-Pass Filter

Equation 1 through Equation 2 show calculations for corner frequency and gain:

$$f_{-3 dB} = \frac{1}{2\pi R_1 C_1}$$

$$\frac{V_{OUT}}{V_{IN}} = \left(1 + \frac{R_F}{R_G}\right) \left(\frac{1}{1 + sR_1 C_1}\right)$$
(2)

### **Typical Applications (continued)**

#### 8.2.1.1 Design Requirements

When receiving low-level signals, limiting the bandwidth of the incoming signals into the system is often required. The simplest way to establish this limited bandwidth is to place an RC filter at the noninverting terminal of the amplifier, as shown in Figure 29. If a steeper attenuation level is required, a two-pole or higher-order filter may be used.

#### 8.2.1.2 Detailed Design Procedure

The design goals for this circuit include these parameters:

- A noninverting gain of 10 V/V (20 dB)
- Design a single-pole response circuit with –3-dB rolloff at 15.9 kHz and 159 Hz
- Modify the design to increase attenuation level to -40 dB/decade (Sallen-Key Filter)

Use these design values:

- $C_1 = 0 \text{ nF}, 10 \text{ nF}, 1 \mu\text{F}$
- $R_1 = 1 k\Omega$
- $R_G = 10 \text{ k}\Omega$
- $R_F = 90 \text{ k}\Omega$

Figure 30 shows how the output voltage of OPA340 changes over frequency depending on the value of  $C_1$  with a constant  $R_1$  of 1 k $\Omega$ . Without any filtering of the input signal ( $C_1$  = 0), the -3-dB effective bandwidth is a function of the OPA340 unity-gain bandwidth and closed-loop gain,  $f_{(-3dB)}$  = UGBW/A<sub>CL</sub>, where A<sub>CL</sub> is closed-loop gain and UGBW denotes unity-gain bandwidth. Thus, for a closed-loop gain = 10,  $f_{(-3dB)}$  = 1 MHz/10 =100 kHz; see Figure 30.

To further limit the output bandwidth, an appropriate choice of  $C_1$  must be made: for  $C_1$  = 10 nF,  $f_C = \frac{1}{2\pi \times R_1 C_1} = \frac{1}{2\pi \times 1^3 \times 1^{-8}} = 15.9 \text{ kHz}.$ 

To further limit the bandwidth, a larger  $C_1$  must be used: choosing  $C_1 = 1 \mu F$ ,

$$f_C = \frac{1}{2\pi \times R_1 C_1} = \frac{1}{2\pi \times 1^3 \times 1^{-6}} = 159 \text{ Hz (see Figure 30)}.$$

### 8.2.1.3 Application Curve

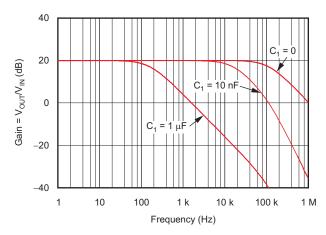


Figure 30. OPA340 Single-Pole AC Gain vs Frequency Response

#### Typical Applications (continued)

#### 8.2.2 Two-Pole, Low-Pass Filter

If even more attenuation is required, a multiple pole filter is required. The Sallen-Key filter may be used for this task, as shown in Figure 31. For best results, the amplifier must have effective bandwidth that is at least 10 times higher than the filter cutoff frequency. Failure to follow this guideline results in a phase shift of the amplifier, which in turn leads to lower precision of the filter bandwidth. Additionally, to minimize the loading effect between multiple RC pairs on overall the filter cutoff frequency, choose  $R = 10 \times R_1$  and  $C_2 = C1/10$ ; see Figure 32.

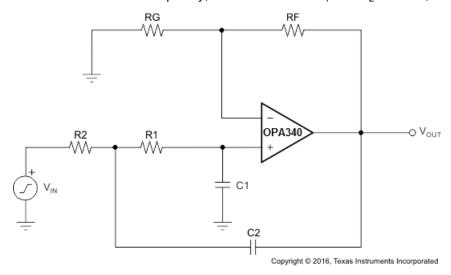


Figure 31. Two-Pole, Lowpass Filter

Equation 3 through Equation 5 show calculations for corner frequency and gain:

$$f_{\rm C} = \frac{1}{2\pi\sqrt{R_1C_1R_2C_2}}$$
 (3)

$$\frac{V_{OUT(s)}}{V_{IN(s)}} = \frac{G(2\pi f_c)^2}{s^2 + 2\zeta(2\pi f_c)s + (2\pi f_c)^2}$$
(4)

$$G = \frac{R_G + R_F}{R_G} \tag{5}$$

#### 8.2.2.1 Detailed Design Procedure

Use these design values:

- C<sub>1</sub> = 10 nF and C<sub>2</sub> = 1 nF
- $R_1 = 1 \text{ k}\Omega$  and  $R_2 = 10 \text{ k}\Omega$
- $R_G = 10 \text{ k}\Omega$
- $R_F = 90 \text{ k}\Omega$

Figure 32 shows the Sallen-Key filter second-order response for different RC values: for R and C values above, 
$$f_{\text{C}} = \frac{1}{2\pi\sqrt{R_1C_1R_2C_2}} = \frac{1}{2\pi\sqrt{1^3\times 1^{-8}\times 1^4\times 1^{-9}}} = 15.9 \text{ kHz}.$$

To further limit the bandwidth, a larger RC value must be used: increasing C values 100 times, such as  $C_1 = 1 \mu F$  and  $C_2 = 0.1 \mu F$ , with unchanged resistors, results in the second-order rolloff at  $f_C = \frac{1}{2\pi\sqrt{R_1C_1R_2C_2}} = \frac{1}{2\pi\sqrt{1^3 \times 1^{-6} \times 1^4 \times 1^{-7}}} = 159 \text{ Hz. See Figure 32.}$ 

### **Typical Applications (continued)**

### 8.2.2.2 Application Curve

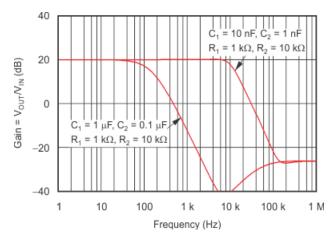


Figure 32. OPA340 Two-Pole, Lowpass Sallen-Key AC Gain vs Frequency Response

### 9 Power Supply Recommendations

The OPAx340 is specified for operation from 2.7 V to 5.5 V (±1.35 V to ±2.75 V).

#### **CAUTION**

Supply voltages larger than 7 V can permanently damage the device (see the *Absolute Maximum Ratings*).

TI recommends placing 0.1- $\mu$ F bypass capacitors close to the power-supply pins to reduce errors coupling in from noisy or high-impedance power supplies.

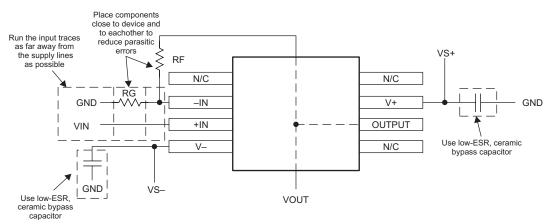
### 10 Layout

### 10.1 Layout Guidelines

Pay attention to good layout practices. Keep traces short and when possible, use a printed-circuit board (PCB) ground plane with surface-mount components placed as close to the device pins as possible. Place a 0.1-μF capacitor closely across the supply pins. Apply these guidelines throughout the analog circuit to improve performance and provide benefits, such as reducing the electromagnetic interference (EMI) susceptibility.

Operational amplifiers vary in susceptibility to radio frequency interference (RFI). RFI can generally be identified as a variation in offset voltage or DC signal levels with changes in the interfering RF signal. The OPA340 is specifically designed to minimize susceptibility to RFI and demonstrates remarkably low sensitivity compared to previous generation devices. Strong RF fields can still cause varying offset levels.

#### 10.2 Layout Example



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Figure 33. Layout Recommendation

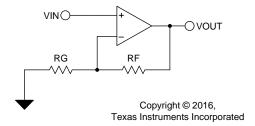


Figure 34. Schematic Representation

### 11 Device and Documentation Support

### 11.1 Device Support

#### 11.1.1 Development Support

#### 11.1.1.1 TINA-TI™ (Free Software Download)

TINA™ is a simple, powerful, and easy-to-use circuit simulation program based on a SPICE engine. TINA-TI™ is a free, fully-functional version of the TINA software, preloaded with a library of macro models in addition to a range of both passive and active models. TINA-TI provides all the conventional DC, transient, and frequency domain analysis of SPICE, as well as additional design capabilities.

Available as a free download from the Analog eLab Design Center, TINA-TI offers extensive post-processing capability that allows users to format results in a variety of ways. Virtual instruments offer the ability to select input waveforms and probe circuit nodes, voltages, and waveforms, creating a dynamic guick-start tool.

#### NOTE

These files require that either the TINA software (from DesignSoft™) or TINA-TI software be installed. Download the free TINA-TI software from the TINA-TI folder.

#### 11.1.1.2 DIP Adapter EVM

The DIP Adapter EVM tool provides an easy, low-cost way to prototype small surface mount ICs. The evaluation tool these TI packages: D or U (SOIC-8), PW (TSSOP-8), DGK (MSOP-8), DBV (SOT23-6, SOT23-5 and SOT23-3), DCK (SC70-6 and SC70-5), and DRL (SOT563-6). The DIP Adapter EVM may also be used with terminal strips or may be wired directly to existing circuits.

#### 11.1.1.3 Universal Operational Amplifier EVM

The Universal Op Amp EVM is a series of general-purpose, blank circuit boards that simplify prototyping circuits for a variety of IC package types. The evaluation module board design allows many different circuits to be constructed easily and quickly. Five models are offered, with each model intended for a specific package type. PDIP, SOIC, MSOP, TSSOP and SOT23 packages are all supported.

#### NOTE

These boards are unpopulated, so users must provide their own ICs. TI recommends requesting several operational amplifier device samples when ordering the Universal Op Amp EVM.

#### 11.1.1.4 TI Precision Designs

TI Precision Designs are analog solutions created by Ti's precision analog applications experts and offer the theory of operation, component selection, simulation, complete PCB schematic and layout, bill of materials, and measured performance of many useful circuits. TI Precision Designs are available online at <a href="http://www.ti.com/ww/en/analog/precision-designs/">http://www.ti.com/ww/en/analog/precision-designs/</a>.

#### 11.1.1.5 WEBENCH® Filter Designer

WEBENCH® Filter Designer is a simple, powerful, and easy-to-use active filter design program. The WEBENCH Filter Designer lets you create optimized filter designs using a selection of TI operational amplifiers and passive components from TI's vendor partners.

Available as a web-based tool from the WEBENCH® Design Center, WEBENCH® Filter Designer allows you to design, optimize, and simulate complete multistage active filter solutions within minutes.

#### 11.2 Documentation Support

#### 11.2.1 Related Documentation

For related documentation, see the following application reports and publications (available for download from www.ti.com):

- 12-Bit High Speed Micro Power Sampling Analog-To-Digital Converter (SBAS061)
- A Dual-Polarity, Bidirectional Current-Shunt Monitor (SLYT311)
- OPA340, OPA2340, OPA4340 EMI Immunity Performance (SBOZ010)
- Getting the Full Potential from your ADC (SBAA069)
- Feedback Plots Define Op Amp AC Performance (SBOA015)
- Capacitive Load Drive Solution Using an Isolation Resistor (TIPD128)
- Circuit Board Layout Techniques (SLOA089)

#### 11.3 Related Links

Table 1 lists quick access links. Categories include technical documents, support and community resources, tools and software, and quick access to sample or buy.

**TECHNICAL SUPPORT & TOOLS & PARTS** PRODUCT FOLDER **SAMPLE & BUY DOCUMENTS SOFTWARE** COMMUNITY **OPA340** Click here Click here Click here Click here Click here **OPA2340** Click here Click here Click here Click here Click here **OPA4340** Click here Click here Click here Click here Click here

Table 1. Related Links

### 11.4 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. In the upper right corner, click on *Alert me* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

#### 11.5 Community Resources

The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use.

TI E2E™ Online Community TI's Engineer-to-Engineer (E2E) Community. Created to foster collaboration among engineers. At e2e.ti.com, you can ask questions, share knowledge, explore ideas and help solve problems with fellow engineers.

**Design Support** *TI's Design Support* Quickly find helpful E2E forums along with design support tools and contact information for technical support.

#### 11.6 Trademarks

MicroAmplifier, TINA-TI, E2E are trademarks of Texas Instruments.

WEBENCH is a registered trademark of Texas Instruments.

TINA, DesignSoft are trademarks of DesignSoft, Inc.

All other trademarks are the property of their respective owners.

### 11.7 Electrostatic Discharge Caution



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

### 11.8 Glossary

SLYZ022 — TI Glossary.

This glossary lists and explains terms, acronyms, and definitions.

# 12 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

# PACKAGE OPTION ADDENDUM

13-Aug-2021

### **PACKAGING INFORMATION**

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead finish/ Ball material	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
OPA2340EA/250	ACTIVE	VSSOP	DGK	8	250	RoHS & Green	NIPDAUAG	Level-2-260C-1 YEAR		A40A	Samples
OPA2340EA/250G4	ACTIVE	VSSOP	DGK	8	250	RoHS & Green	NIPDAUAG	Level-2-260C-1 YEAR		A40A	Samples
OPA2340EA/2K5	ACTIVE	VSSOP	DGK	8	2500	RoHS & Green	NIPDAUAG	Level-2-260C-1 YEAR		A40A	Samples
OPA2340EA/2K5G4	ACTIVE	VSSOP	DGK	8	2500	RoHS & Green	NIPDAUAG	Level-2-260C-1 YEAR		A40A	Samples
OPA2340PA	ACTIVE	PDIP	Р	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type		OPA2340PA	Samples
OPA2340PAG4	ACTIVE	PDIP	Р	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type		OPA2340PA	Samples
OPA2340UA	ACTIVE	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	OPA 2340UA	Samples
OPA2340UA/2K5	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR		OPA 2340UA	Samples
OPA2340UA/2K5G4	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR		OPA 2340UA	Samples
OPA340NA/250	ACTIVE	SOT-23	DBV	5	250	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	A40	Samples
OPA340NA/250G4	ACTIVE	SOT-23	DBV	5	250	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	A40	Samples
OPA340NA/3K	ACTIVE	SOT-23	DBV	5	3000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	A40	Samples
OPA340NA/3KG4	ACTIVE	SOT-23	DBV	5	3000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	A40	Samples
OPA340PA	ACTIVE	PDIP	Р	8	50	RoHS & Green	NIPDAU	N / A for Pkg Type	-40 to 85	OPA340PA	Samples
OPA340UA	ACTIVE	SOIC	D	8	75	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	OPA 340UA	Samples
OPA340UA/2K5	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	OPA 340UA	Samples
OPA340UA/2K5G4	ACTIVE	SOIC	D	8	2500	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	OPA 340UA	Samples
OPA4340EA/250	ACTIVE	SSOP	DBQ	16	250	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	OPA 4340EA	Samples

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead finish/ Ball material	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
OPA4340EA/250G4	ACTIVE	SSOP	DBQ	16	250	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	OPA 4340EA	Samples
OPA4340EA/2K5	ACTIVE	SSOP	DBQ	16	2500	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	OPA 4340EA	Samples
OPA4340UA	ACTIVE	SOIC	D	14	50	RoHS & Green	NIPDAU	Level-3-260C-168 HR	-40 to 85	OPA4340UA	Samples
OPA4340UA/2K5	ACTIVE	SOIC	D	14	2500	RoHS & Green	NIPDAU	Level-3-260C-168 HR	-40 to 85	OPA4340UA	Samples
OPA4340UA/2K5G4	ACTIVE	SOIC	D	14	2500	RoHS & Green	NIPDAU	Level-3-260C-168 HR	-40 to 85	OPA4340UA	Samples
OPA4340UAG4	ACTIVE	SOIC	D	14	50	RoHS & Green	NIPDAU	Level-3-260C-168 HR	-40 to 85	OPA4340UA	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

- (3) MSL, Peak Temp. The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
- (6) Lead finish/Ball material Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

### **PACKAGE OPTION ADDENDUM**

13-Aug-2021

**Important Information and Disclaimer:** The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

#### OTHER QUALIFIED VERSIONS OF OPA340:

■ Enhanced Product : OPA340-EP

NOTE: Qualified Version Definitions:

• Enhanced Product - Supports Defense, Aerospace and Medical Applications

### TAPE AND REEL INFORMATION





	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

### QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



#### \*All dimensions are nominal

^All dimensions are nominal	1	1			1			1		1		
Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
OPA2340EA/250	VSSOP	DGK	8	250	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
OPA2340EA/2K5	VSSOP	DGK	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
OPA2340UA/2K5	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
OPA340NA/250	SOT-23	DBV	5	250	179.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
OPA340NA/250	SOT-23	DBV	5	250	178.0	9.0	3.3	3.2	1.4	4.0	8.0	Q3
OPA340NA/3K	SOT-23	DBV	5	3000	178.0	9.0	3.3	3.2	1.4	4.0	8.0	Q3
OPA340NA/3K	SOT-23	DBV	5	3000	179.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
OPA340UA/2K5	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
OPA4340EA/250	SSOP	DBQ	16	250	180.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
OPA4340EA/2K5	SSOP	DBQ	16	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
OPA4340UA/2K5	SOIC	D	14	2500	330.0	16.4	6.5	9.0	2.1	8.0	16.0	Q1

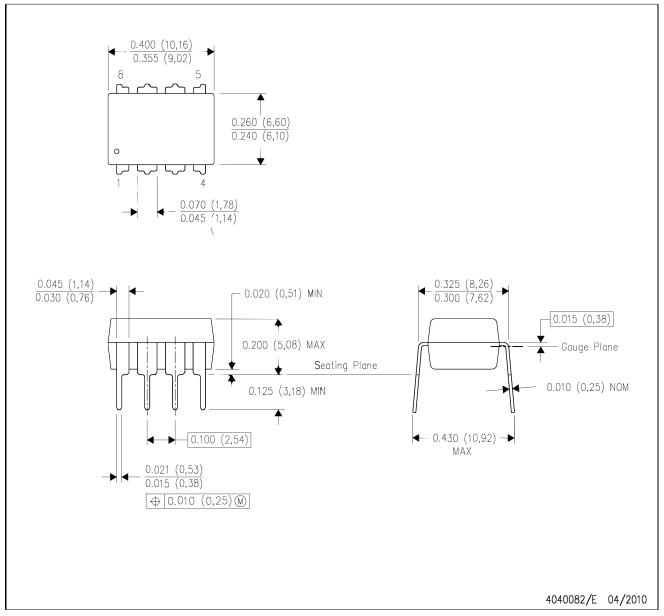


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
OPA2340EA/250	VSSOP	DGK	8	250	366.0	364.0	50.0
OPA2340EA/2K5	VSSOP	DGK	8	2500	366.0	364.0	50.0
OPA2340UA/2K5	SOIC	D	8	2500	853.0	449.0	35.0
OPA340NA/250	SOT-23	DBV	5	250	213.0	191.0	35.0
OPA340NA/250	SOT-23	DBV	5	250	180.0	180.0	18.0
OPA340NA/3K	SOT-23	DBV	5	3000	180.0	180.0	18.0
OPA340NA/3K	SOT-23	DBV	5	3000	213.0	191.0	35.0
OPA340UA/2K5	SOIC	D	8	2500	853.0	449.0	35.0
OPA4340EA/250	SSOP	DBQ	16	250	210.0	185.0	35.0
OPA4340EA/2K5	SSOP	DBQ	16	2500	853.0	449.0	35.0
OPA4340UA/2K5	SOIC	D	14	2500	853.0	449.0	35.0

# P (R-PDIP-T8)

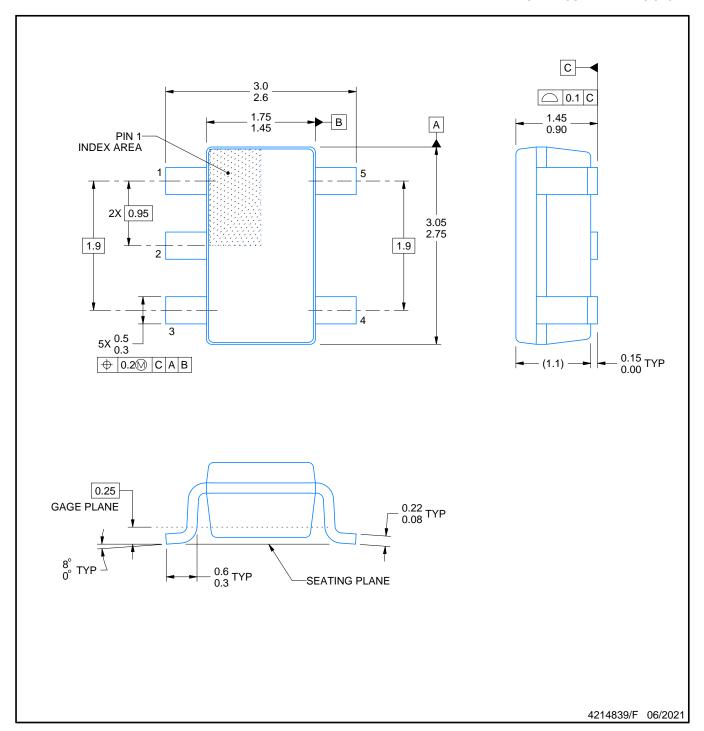
# PLASTIC DUAL-IN-LINE PACKAGE



- A. All linear dimensions are in inches (millimeters).
- B. This drawing is subject to change without notice.
- C. Falls within JEDEC MS-001 variation BA.



SMALL OUTLINE TRANSISTOR



- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.

  2. This drawing is subject to change without notice.

  3. Reference JEDEC MO-178.

- 4. Body dimensions do not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.25 mm per side.

SMALL OUTLINE TRANSISTOR



NOTES: (continued)

5. Publication IPC-7351 may have alternate designs.6. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

SMALL OUTLINE TRANSISTOR



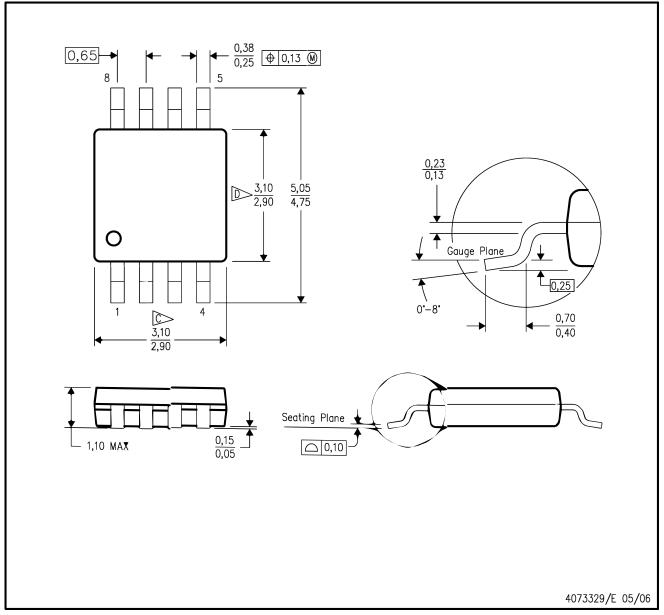
NOTES: (continued)

<sup>7.</sup> Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

<sup>8.</sup> Board assembly site may have different recommendations for stencil design.

# DGK (S-PDSO-G8)

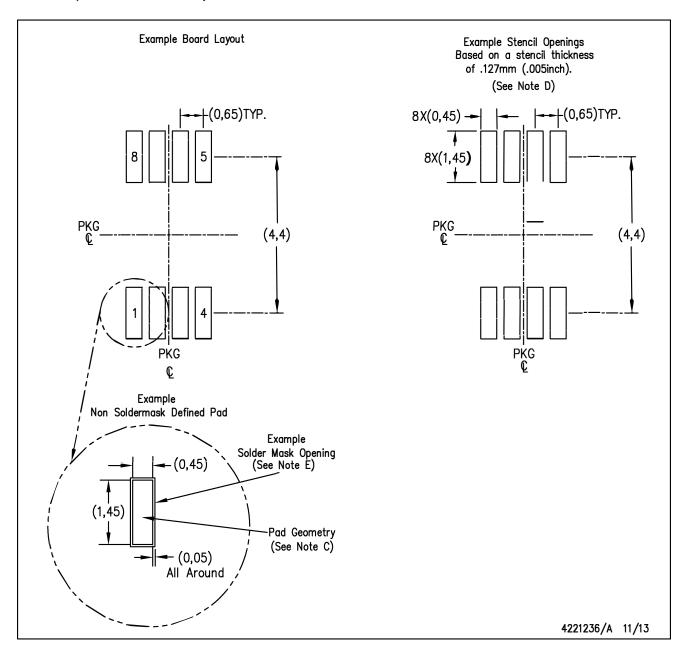
# PLASTIC SMALL-OUTLINE PACKAGE



- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 per end.
- Body width does not include interlead flash. Interlead flash shall not exceed 0.50 per side.
- E. Falls within JEDEC MO-187 variation AA, except interlead flash.

# DGK (S-PDSO-G8)

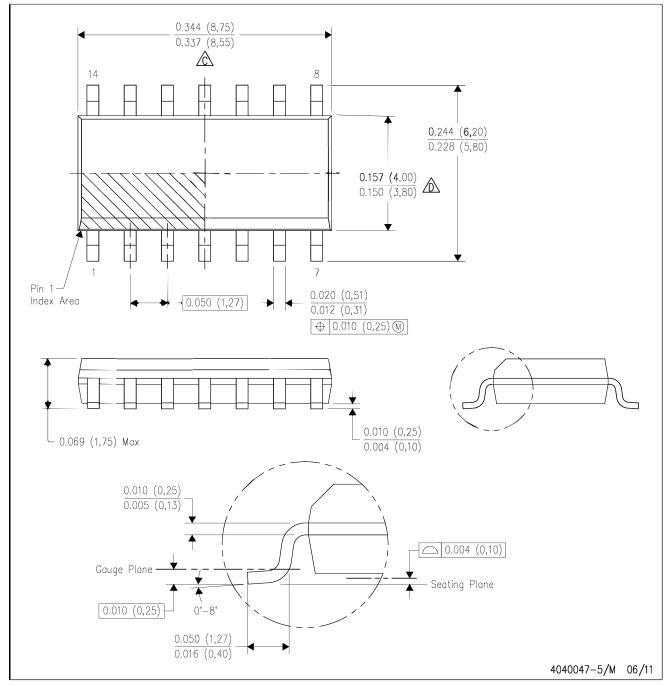
### PLASTIC SMALL OUTLINE PACKAGE



- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
- E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.

# D (R-PDS0-G14)

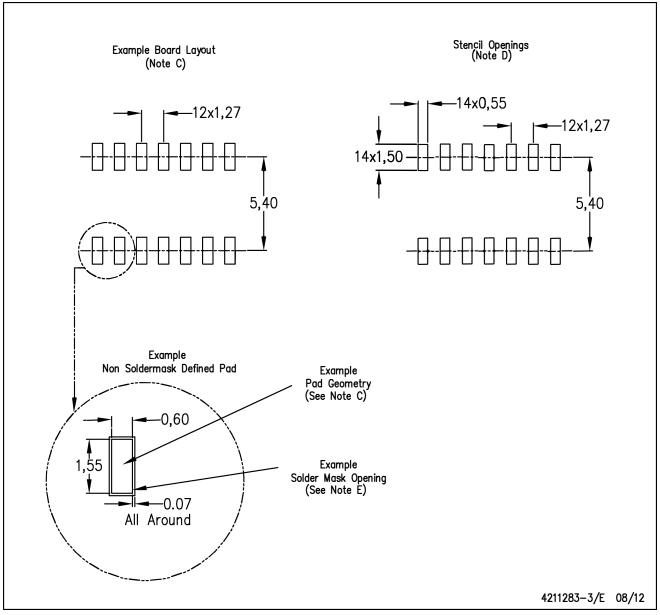
### PLASTIC SMALL OUTLINE



- A. All linear dimensions are in inches (millimeters).
- B. This drawing is subject to change without notice.
- Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.006 (0,15) each side.
- Body width does not include interlead flash. Interlead flash shall not exceed 0.017 (0,43) each side.
- E. Reference JEDEC MS-012 variation AB.

# D (R-PDSO-G14)

# PLASTIC SMALL OUTLINE



- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
- E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.

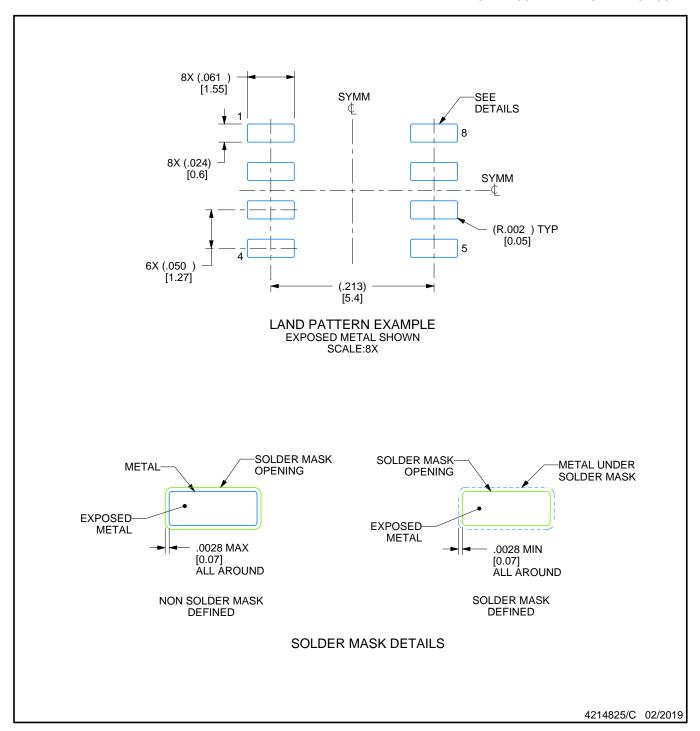


SMALL OUTLINE INTEGRATED CIRCUIT



- 1. Linear dimensions are in inches [millimeters]. Dimensions in parenthesis are for reference only. Controlling dimensions are in inches. Dimensioning and tolerancing per ASME Y14.5M.
- 2. This drawing is subject to change without notice.
- 3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed .006 [0.15] per side.
- 4. This dimension does not include interlead flash.
- 5. Reference JEDEC registration MS-012, variation AA.

SMALL OUTLINE INTEGRATED CIRCUIT



NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

SMALL OUTLINE INTEGRATED CIRCUIT

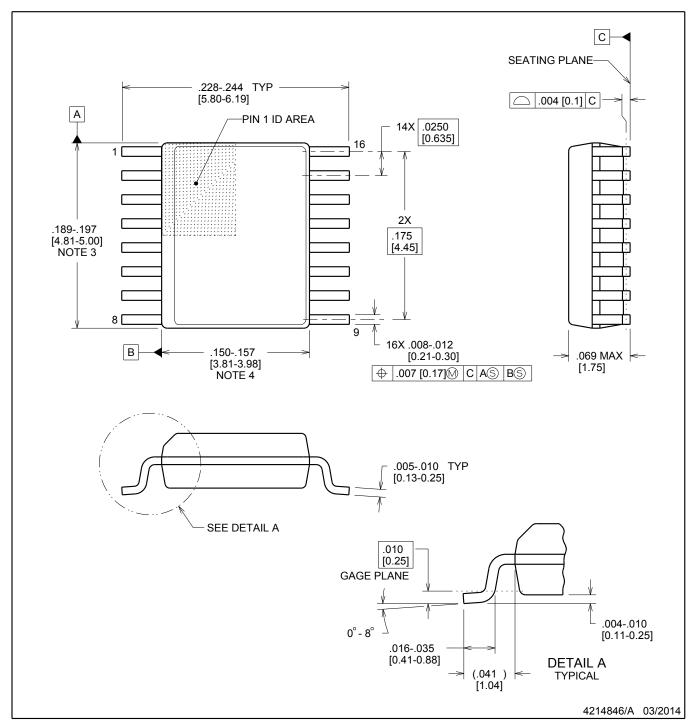


NOTES: (continued)

- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.

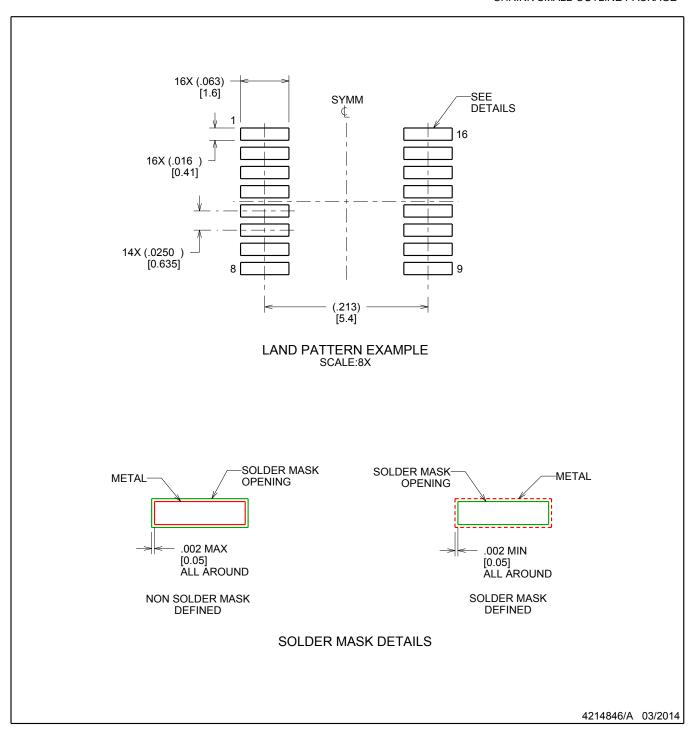


SHRINK SMALL-OUTLINE PACKAGE



- 1. Linear dimensions are in inches [millimeters]. Dimensions in parenthesis are for reference only. Controlling dimensions are in inches. Dimensioning and tolerancing per ASME Y14.5M.
- 2. This drawing is subject to change without notice.
- 3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed .006 inch, per side.
- 4. This dimension does not include interlead flash.5. Reference JEDEC registration MO-137, variation AB.

SHRINK SMALL-OUTLINE PACKAGE

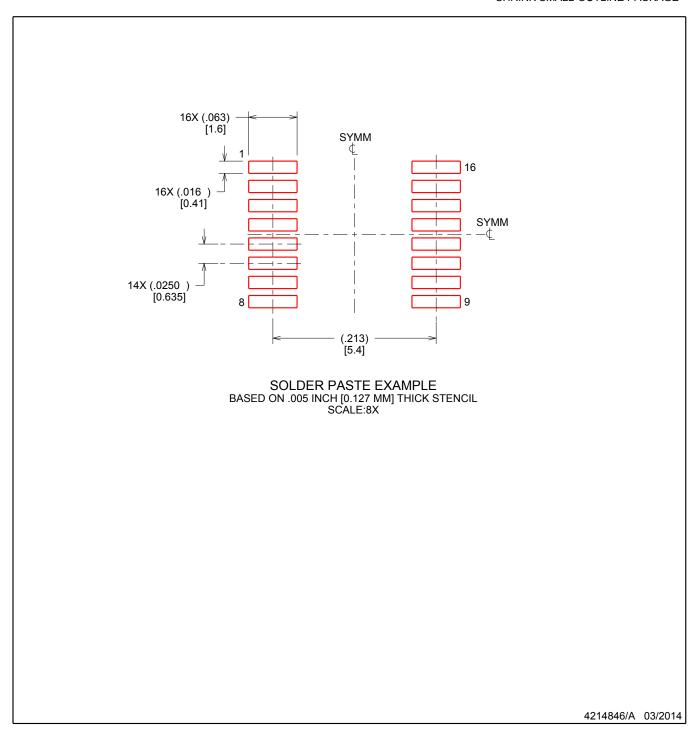


NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

SHRINK SMALL-OUTLINE PACKAGE



NOTES: (continued)

- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.